Nonlinear Energy-Maximizing Optimal Control of Wave Energy Systems: A Moment-Based Approach

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Abstract—Linear dynamics are virtually always assumed when designing optimal controllers for wave energy converters (WECs), motivated by both their simplicity and computational convenience. Nevertheless, unlike traditional tracking control applications, the assumptions under which the linearization of WEC models is performed are challenged by the energy-maximizing controller itself, which intrinsically enhances device motion to maximize power extraction from incoming ocean waves. In this article, we present a moment-based energy-maximizing control strategy for WECs subject to nonlinear dynamics. We develop a framework under which the objective function (and system variables) can be mapped to a finite-dimensional tractable nonlinear program, which can be efficiently solved using stateof-the-art nonlinear programming solvers. Moreover, we show that the objective function belongs to a class of generalized convex functions when mapped to the moment domain, guaranteeing the existence of a global energy-maximizing solution and giving explicit conditions for when a local solution is, effectively, a global maximizer. The performance of the strategy is demonstrated through a case study, where we consider (state and input-constrained) energy maximization for a state-of-theart CorPower-like WEC, subject to different hydrodynamic nonlinearities.

Index Terms—Energy maximization, moment, moment matching, nonlinear optimal control, wave energy.

I. INTRODUCTION

E NERGY-MAXIMIZING control of wave energy converters (WECs) has been shown to be one of the fundamental contributions toward efficient energy extraction from ocean

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waves [1], [2]. In particular, optimal control strategies significantly improve maximum time-averaged power extraction from waves, potentially reducing the levelized cost of energy and hence contributing to the roadmap toward successful commercialization of WECs.

Any energy-maximizing control strategy for WECs must consider the inherent physical limitations of both the device itself and the power take-off (PTO) actuator dynamics such that energy extraction is maximized while also minimizing the risk of component damage. The above control specifications form the basis for a range of studies, which are mostly formulated in terms of an optimal control problem (OCP) subject to both state and input constraints. A comprehensive review of these strategies can be found in [2] and [3].

Linear dynamics are virtually always considered when designing optimal controllers for WECs (see [3, Table 3]), motivated by both their simplicity (in terms of formulation and solution of the corresponding OCP), and their associated computational convenience. In other words, these model-based control strategies must be computed in real time,¹ therefore limiting the computational complexity of the hydrodynamic models employed. Moreover, there is also a limit to the complexity of mathematical models for which an optimal control solution can be effectively found, either algebraically or numerically. Another strongly contributing factor to the use of linear dynamics is that linear hydrodynamic theory is a well-established field where considerable effort and refinement has been expended in the calculation of linear hydrodynamic parameters. There is therefore little appetite to extend these models to include nonlinear effects.

Nevertheless, despite the list of motives described above, the linearity assumption has been recently an object of debate (see [6], [7]); WECs are, by their nature, prone to show significant and diverse nonlinear effects since their principal aim, pursued by the optimal control strategy, is to enhance the amplitude of motion to maximize power extraction. In other words, and in contrast to traditional set-point tracking control systems, where the control system ensures that the system operation is around the set point and actively attempts to reduce the variance around this point, the assumptions under which the linearization of WEC models is performed are

¹Though beyond the scope of this study, we note that control techniques that optimize control parameters/laws offline have also been proposed within the WEC literature. The reader is referred to, for instance [4] and [5].

1063-6536 © 2021 IEEE. Personal use is permitted, but republication/redistribution requires IEEE permission. See https://www.ieee.org/publications/rights/index.html for more information. challenged by the controller itself, particularly in relation to small movements around the equilibrium position. This may, in certain conditions, return poor results, both in terms of accuracy of motion prediction that in terms of power production assessment [6], which are the key variables involved in any energy-maximizing OCP.

This provides significant motivation for optimal control strategies that can effectively handle nonlinear effects, both in terms of the well-posedness of the OCP (i.e., existence of global energy-maximizing solutions) and realtime capabilities. Some attempts to address this problem have already been offered within the WEC control literature. We provide a brief review in the following paragraphs.

The papers [8] and [9] propose one of the first nonlinear optimal control formulations in the WEC literature: nonlinear energy-maximizing model predictive control (MPC) applied to (different) heaving point absorbers. The nonlinear effects considered in [8] arise from the forces exerted by the mooring system utilized to attach the device to the seafloor. Nevertheless, nonlinear hydrodynamic effects, such as viscous forces (which are significant in this type of device [7]) are disregarded, while (nonparametric) radiation forces are completely neglected. Concerning computational efficiency, the authors specifically claim that the study in [8] does not focus on real-time applicability, but rather on highlighting the advantages of using nonlinear MPC compared to its linear counterpart. A follow up of this study can be found in [10], where the discretization for the MPC is modified (from a zeroto a second-order hold), and a state estimator is incorporated. On a different note, while Tom and Yeung [9] assumed linear dynamics for the WEC, they compute the control force using a nonlinear MPC formulation given by a time-varying system (written in the explicit form), which naturally induces a timevarying objective function.

The papers [11]–[13] apply pseudospectral optimal control methods (see [14] for a review) to different WEC technologies, subject to a number of nonlinearities: viscous forces (see [11], [12]), nonlinear hydrostatic restoring force (see [12], [13]), and a nonideal PTO system ([11]). Although these pseudospectral formulations are indeed appealing from a computational perspective, the results in [11]–[13] do not guarantee existence of solutions to the proposed OCPs. Moreover, even if a solution is found, it is not clear under which conditions this energy-maximizing control law is effectively a globally optimal solution.

Recently, a novel energy-maximizing optimal control framework for WECs was presented in [15] and [16]. This approach is based on the concept of moment (discussed in Section II) and maps the original energy-maximizing OCP into a concave quadratic program (QP), systematically guaranteeing a unique solution for the original energy-maximizing control objective, subject to both state and input constraints. Nevertheless, the mathematical framework in [15] and [16] considers WECs subject to linear dynamics, hence hindering the application of this strategy to more realistic scenarios in which nonlinear effects become significant, as previously discussed in this section. Exploiting the concept of nonlinear moment introduced in [17], this article presents an energy-maximizing control strategy for WECs subject to nonlinear dynamics. In particular, we propose an approximation method that allows us to efficiently parameterize the energy-maximizing OCP subject to both state and input constraints. To this end, this article provides the following contributions.

- 1) We propose a method to map the objective function (and system variables) to a finite-dimensional tractable nonlinear program (NP), which can be efficiently solved using state-of-the-art nonlinear programming solvers (see [18]).
- 2) By showing that the objective function arising from the proposed moment-based strategy belongs to a family of approximately convex/concave mappings (particularly to the so-called outer Γ -convex/concave [19] functions), we guarantee the existence of a global energy-maximizing solution, under mild assumptions.
- 3) In analogy to the case of convex/concave functions, where each local solution is also global, we give explicit conditions to determine whether a local energymaximizing solution is, effectively, a global maximizer for the proposed moment-based OCP, having strong practical implications when numerically solving the associated NP.
- Finally, we present an extensive case study based on the energy-maximization problem for a state-ofthe-art CorPower-like WEC subject to hydrodynamic nonlinearities.

The remainder of this article is organized as follows. Section II recalls the key theoretical concepts behind the nonlinear moment-based framework. Section III describes and formalizes the energy-maximizing problem for WECs subject to nonlinear dynamics, while Section IV details the main contributions of this article, i.e., the proposed nonlinear momentbased optimal control strategy. Finally, Section V discusses the application of this approach to a state-of-the-art CorPowerlike heaving WEC, while Section VI encompasses the main conclusions of this study.

A. Notation and Preliminaries

Standard notation is considered throughout this study. \mathbb{R}^+ (\mathbb{R}^-) denotes the set of nonnegative (nonpositive) real numbers. \mathbb{C}^0 denotes the set of pure-imaginary complex numbers. The symbol 0 stands for any zero element, dimensioned according to the context. The notation \mathbb{N}_q indicates the set of all positive natural numbers up to q, i.e., $\mathbb{N}_q = \{1, 2, \dots, q\}$. The symbol \mathbb{I}_n denotes the identity matrix of order *n*, while the notation $\mathbf{1}_{n \times m}$ is used to denote an $n \times m$ Hadamard identity matrix (i.e., an $n \times m$ matrix with all its entries equal to 1). The spectrum of a matrix $A \in \mathbb{R}^{n \times n}$, i.e., the set of its eigenvalues, is denoted by $\lambda(A)$. The superscript \neg denotes the transposition operator. The symbol \bigoplus denotes the direct sum of *n* matrices, i.e., $\bigoplus_{i=1}^{n} A_i = \text{diag}(A_1, A_2, \dots, A_n)$. The notations $\Re\{z\}$ and $\Im\{z\}$, with $z \in \mathbb{C}$, stand for the real- and imaginarypart operators, respectively. The symmetric part of a matrix $A \in \mathbb{R}^{n \times n}$ is defined (and denoted) as $\mathscr{H}\{A\} = (A + A^{\mathsf{T}})/2$.

The Frobenius norm of a matrix $X \in \mathbb{R}^{n \times m}$ is denoted as $||X||_F$, whereas the Euclidean norm of a row vector $V \in \mathbb{R}^n$ is denoted as $||V||_2$. The generalized Dirac- δ function, shifted by $t_j \in \mathbb{R}^+$, is denoted as $\delta_{t_j} = \delta(t - t_j)$. The Kronecker product between two matrices $M_1 \in \mathbb{R}^{n \times m}$ and $M_2 \in \mathbb{R}^{p \times q}$ is denoted by $M_1 \otimes M_2 \in \mathbb{R}^{np \times mq}$. The convolution between two functions f and g over the set $\Omega \subset \mathbb{R}$, i.e., $\int_{\Omega} f(\tau)g(t-\tau)d\tau$ is denoted as f * g. Let f and g be two functions belonging to the Hilbert space $L^2(\Xi) = \{f : \Xi \to \mathbb{R} \mid \int_{\Xi} |f(\tau)|^2 d\tau <$ $+\infty$ }, where $\Xi \subset \mathbb{R}$ is closed. Then, the inner product between f and g is given by $\langle f, g \rangle = \int_{\Xi} f(\tau)g(\tau)d\tau$. The span of the set $\mathscr{X} = \{x_i\}_{i=1}^k \subset \mathscr{Z}$, where \mathscr{Z} is a vector space over the field \mathbb{R} , is denoted as span{ \mathscr{X} }. The closed ball contained in \mathbb{R}^n , with center $z \in \mathbb{R}^n$ and radius $r \in \mathbb{R}^+$, is defined as $\mathscr{B}(z,r) = \{y \in \mathbb{R}^n \mid ||y - z||_2 \le r\}$. Finally, the symbol $\varepsilon_n \in \mathbb{R}^n$ denotes a vector with odd entries equal to 1 and even entries equal to 0.

In the remainder of this section, the formal definitions of two important operators are presented since their definition in the literature can often be ambiguous.

Definition 1 (Kronecker Sum) [20]: The Kronecker sum of two matrices P_1 and P_2 , with $P_1 \in \mathbb{R}^{n \times n}$ and $P_2 \in \mathbb{R}^{k \times k}$, is defined (and denoted) as

$$P_1 \oplus P_2 \triangleq P_1 \otimes \mathbb{I}_k + \mathbb{I}_n \otimes P_2. \tag{1}$$

Definition 2 (Vec Operator) [20]: Given a matrix $P = [p_1, p_2, ..., p_m] \in \mathbb{R}^{n \times m}$, where $p_j \in \mathbb{R}^n$, $j \in \mathbb{N}_m$, the vector valued operator vec is defined as

$$\operatorname{vec}\{P\} \triangleq \begin{bmatrix} p_1 \\ p_2 \\ \vdots \\ p_m \end{bmatrix} \in \mathbb{R}^{nm}.$$
 (2)

Finally, we recall a useful property of the vec operator. *Property 1* [20]: Let $P_3 \in \mathbb{R}^{n \times m}$ and $P_4 \in \mathbb{R}^{m \times q}$. Then

$$\operatorname{vec}\{P_3P_4\} = (\mathbb{I}_q \otimes P_3)\operatorname{vec}\{P_4\} = (P_4^{\mathsf{T}} \otimes \mathbb{I}_n)\operatorname{vec}\{P_3\}.$$
(3)

II. MOMENT-BASED THEORY

To keep this article reasonably self-contained, this section briefly recalls some of the key concepts behind the momentbased framework, as developed in [17] and [21]. In particular (after recalling a set of standing assumptions), this section is mainly concerned with the definition of moment for nonlinear systems.

We consider a nonlinear, single-input-single-output, continuous-time, system described, for $t \in \mathbb{R}^+$, by the set of equations²

$$\dot{x} = f(x, u)$$

$$y = h(x)$$
(4)

with $x(t) \in \mathbb{R}^n$, $u(t) \in \mathbb{R}$, $y(t) \in \mathbb{R}$, and $f : \mathbb{R}^n \times \mathbb{R} \to \mathbb{R}^n$ and $h : \mathbb{R}^n \to \mathbb{R}$ sufficiently smooth mappings defined in the neighborhood of the origin of \mathbb{R}^n . Assume that the origin is an equilibrium point of (4), i.e., f(0, 0) = 0 and h(0) = 0.

²From now on, we drop the dependence on t when clear from the context.

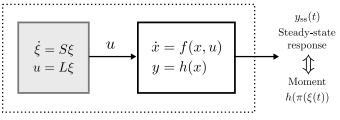


Fig. 1. Block diagram of the composite system (6), showing the relationship between the steady-state output and the moment.

Now, consider a signal generator (often also referred to as exogenous system [22]) described by the equations

$$\begin{split} \dot{\xi} &= S\xi \\ u &= L\xi \end{split} \tag{5}$$

with $\xi(t) \in \mathbb{R}^{\nu}$, $S \in \mathbb{R}^{\nu \times \nu}$, and $L^{\mathsf{T}} \in \mathbb{R}^{\nu}$, and the so-called composite (or interconnected) system

$$\begin{aligned} \dot{\xi} &= S\xi \\ \dot{x} &= f(x, L\xi) \\ y &= h(x). \end{aligned} \tag{6}$$

Following [17] and [21], we now introduce a set of standing assumptions to formalize the definition of moment. Recall that a system is minimal if it is observable, accessible, and described by analytic mappings (see [21, Ch. 2]).

Assumption 1: There exists a mapping π , locally³ defined in a neighborhood of $\xi = 0$ and with $\pi(0) = 0$, which solves the partial differential equation

$$\frac{\partial \pi(\xi)}{\partial \xi} S\xi = f(\pi(\xi), L\xi).$$
(7)

Assumption 2: System (4) is minimal and the pair (S, L) is observable.

Definition 3 ([17], [21]): Consider system (4) and the signal generator (5). Suppose that Assumptions 1 and 2 hold. The mapping $h \circ \pi$ is the moment of system (4) at (*S*, *L*).

Finally, we recall a result that connects the definition of moment (as in Definition 3) with the steady-state response of the output of the interconnected system (6) (as in Fig. 1), introducing an additional assumption concerning the dynamics of (5).

Assumption 3: The signal generator (5) is such that all eigenvalues of S are simple and with zero real part. Moreover, the pair $(S, \xi(0))$ is excitable.⁴

Remark 1: For linear systems, excitability is equivalent to reachability, i.e., with $\xi(0)$ playing the role of the input matrix (see [23]).

Theorem 1 ([17], [21]): Consider system (4) and the signal generator (5). Suppose that Assumptions 2 and 3 hold and that the zero equilibrium of the system (4) is locally exponentially stable. Then, Assumption 1 holds, and the moment of system (4), computed along a trajectory $\xi(t)$, coincides with the steady-state response of the output of the interconnected system (6), i.e., $y_{ss}(t) = h(\pi(\xi(t)))$.

³Statements are local, though global versions can be directly derived. ⁴We refer the reader to [23] for the definition of excitability.

III. PROBLEM FORMULATION: OPTIMAL CONTROL FOR WECs

WEC optimal control design entails an energy-maximization criterion, where the objective is to maximize the absorbed energy from ocean waves over a finite time interval $\Xi = [0, T] \subset \mathbb{R}^+$. To be precise, the useful energy absorbed from incoming waves is converted in the PTO system and can be directly computed as the time integral of converted (instantaneous) power, i.e., this energy-maximizing control procedure can be cast as an OCP, with objective function \mathcal{J} defined as

$$\mathcal{J}(u) = \frac{1}{T} \int_{\Xi} u(\tau) \dot{x}(\tau) d\tau$$
(8)

where $\dot{x} : \Xi \to \mathbb{R}$ and $u : \Xi \to \mathbb{R}$ denote the velocity of the device and control (PTO) force, respectively.

We now briefly recall the basics behind control-oriented WEC modeling in Section III-A, to later pose the OCP for wave energy systems subject to the motion (state and input) constraints in Section III-B.

A. WEC Dynamics

We begin this section by recalling (see [24]) well-known facts behind control-oriented WEC modeling. For simplicity, we assume a one-degree-of-freedom (DoF) device and recall that a similar analysis can be carried out for multi-DoF devices following the moment-based multiple-input–multiple-output approach presented in [16].

Remark 2: In what follows, to simplify notation, the terminology associated with a 1-DoF translational device is employed with some minor abuse. This is without loss of generality, and the same analysis can be carried out for a rotational 1-DoF device.

The equation of motion for such a WEC can be expressed in the time domain, in terms of Newton's second law, as

$$m\ddot{x} = \mathcal{F}_r + \mathcal{F}_h^l + \mathcal{F}_e + \mathcal{F}^{nl} - u \tag{9}$$

where *m* is the mass of the buoy, *x* is the device excursion (displacement), \mathcal{F}_e is the wave excitation force (external input), \mathcal{F}_h^l is the linear component of the hydrostatic restoring force, \mathcal{F}_r is the radiation force, and *u* is the exerted control (PTO) force. The mapping \mathcal{F}^{nl} represents the potential sources of nonlinearity depending on *x* and \dot{x} , such as viscous drag force and nonlinear hydrostatic effects (see Section V-A). Note that, in line with the assumptions in Section II, it is assumed that the mapping \mathcal{F}^{nl} is sufficiently smooth.⁵ The linear component of the hydrostatic force can be written as $\mathcal{F}_h^l(t) = -s_h x(t)$, where $s_h = \rho g_r D$ denotes the hydrostatic stiffness, with ρ the water density, *D* the characteristic area of the device, and g_r the gravitational constant. The radiation force \mathcal{F}_r is modeled based on linear potential theory and, using the well-known Cummins' equation [26], can be written as

$$\mathcal{F}_r(t) = -\mu_{\infty} \ddot{x}(t) - \int_{\mathbb{R}^+} k(\tau) \dot{x}(t-\tau) d\tau \qquad (10)$$

where $\mu_{\infty} = \lim_{\omega \to +\infty} \tilde{A}(\omega) > 0$ is the added mass at infinite frequency, $\tilde{A}(\omega)$ is the radiation added mass⁶ and $k : \mathbb{R}^+ \to \mathbb{R}^+$ is the (causal) radiation impulse response function containing the memory effect of the fluid response. Finally, the equation of motion of the WEC is given by

$$\mathcal{M}\ddot{x} + k * \dot{x} + s_h x - \mathcal{F}^{nl} = \mathcal{F}_e - u \tag{11}$$

with $\mathcal{M} = m + \mu_{\infty}$. We note that (11) is of a Volterra integrodifferential form, specifically of the convolution class.⁷

B. Physical Limitations: State and Input Constraints

As it is well known in the wave energy literature, the unconstrained energy-maximizing optimal control law, i.e., the maximizer of \mathcal{J} in (8), often requires unrealistic device motion and excessively high PTO (control) forces, which consigns this optimal unconstrained solution to the academic realm, far from being practically viable. Aiming to derive an implementable solution, we consider constraints on both the displacement and the velocity of the WEC, *x* and \dot{x} , and on the exerted control force *u*. This guarantees that the physical limits associated with device and actuator dynamics are consistently respected while effectively maximizing, at the same time, absorbed energy from incoming waves. This set of constraints can be compactly written as

$$\begin{cases} |x(t)| \le X_{\max} \\ |\dot{x}(t)| \le V_{\max} \\ |u(t)| \le U_{\max} \end{cases}$$
(12)

with $t \in \mathbb{R}^+$, and where $\{X_{\max}, V_{\max}, U_{\max}\} \subset \mathbb{R}^+$. Having knowledge of the control objective function defined in (8) (i.e., the mapping \mathcal{J}), the nonlinear dynamics of the WEC in (11), and the set of state and input constraints defined in (12), the constrained energy-maximizing OCP can be posed as follows.

Problem 1 (Energy-Maximizing OCP): Find an optimal control input $u^{\text{opt}} : \Xi \to \mathbb{R}$ such that

$$u^{\text{opt}} = \arg \max_{u} \mathcal{J}$$

s.t.:
$$\begin{cases} \text{nonlinear WEC dynamics (11)} \\ \text{state and input constraints (12).} \end{cases}$$
 (13)

Remark 3: We assume that the set of state and input constraints in (13) is feasible, i.e., (12) is given in terms of a feasible set (region). We refer the reader to, for instance, [28], for an explicit analysis and discussion of the feasibility problem for WEC energy-maximizing control systems, as a function of the wave excitation input.

IV. NONLINEAR MOMENT-BASED WEC CONTROL

In this section, we present the main contributions of this article. First, we develop a method to find approximate solutions to the problem of determining the moment of nonlinear systems describing WECs. Second, we formulate and solve the unconstrained OCP in the moment domain. Third, we extend the solution to the state and input constrained problem.

⁵If the mapping \mathcal{F}^{nl} is nonsmooth, smooth approximations can be used to apply the framework proposed in our study. Note that smooth approximations have been previously exploited, within the WEC control literature, for example, in [25].

⁶See [24] for the definition of $\tilde{A}(\omega)$.

⁷The interested reader is referred to [16] and [29] for further detail on this class of integrodifferential operators.

A. Approximation of Moments for Nonlinear Systems

We begin by rewriting the equation of motion (11) in state space as

$$\dot{\phi} = f(\phi, \mathscr{U}) = A\phi - B(k * C\phi) + B\mathscr{U} + f_{nl}(\phi)$$

$$y = h(\phi) = C\phi$$
(14)

where $\varphi(t) = [\varphi_1(t), \varphi_2(t)]^{\mathsf{T}} = [x(t), \dot{x}(t)]^{\mathsf{T}} \in \mathbb{R}^2$, is the state vector of the continuous-time model and $y = \dot{x}$ is the output of the system (assuming the velocity as the measurable output of the device). The function $\mathscr{U} : \mathbb{R}^+ \to \mathbb{R}$, defined as the input of system (14), is given by the expression⁸

$$\mathscr{U} = \mathcal{F}_e - u. \tag{15}$$

Under this representation, the triple of matrices (A, B, C) and the nonlinear mapping $f_{nl} : \mathbb{R}^2 \to \mathbb{R}^2$ forming the equation of motion (14) are given by

$$A = \begin{bmatrix} 0 & 1 \\ -s_h \mathcal{M}^{-1} & 0 \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ \mathcal{M}^{-1} \end{bmatrix}, \quad C^{\mathsf{T}} = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$
$$f_{nl}(\varphi) = \begin{bmatrix} 0 \\ \mathcal{F}^{nl}(\varphi) \end{bmatrix}. \tag{16}$$

We recall from [30] that the standard assumption for the mathematical representation of wave excitation forces in ocean engineering applications is that \mathcal{F}_e can be written as a finite sum of harmonics of a so-called fundamental frequency ω_0 . Following the moment-based theory presented in Section II, the wave excitation force input can be written as a signal generator described, for $t \in \mathbb{R}^+$, by the set of equations

$$\dot{\xi} = S\xi$$
$$\mathcal{F}_e = L_e\xi \qquad (17)$$

where $\xi(t) \in \mathbb{R}^{\nu}$, $L_e^{\mathsf{T}} \in \mathbb{R}^{\nu}$ and the dynamic matrix $S \in \mathbb{R}^{\nu \times \nu}$ can be written in a block-diagonal form as

$$S = \bigoplus_{p=1}^{f} \begin{bmatrix} 0 & p\omega_0 \\ -p\omega_0 & 0 \end{bmatrix}$$
(18)

with $v = 2\tilde{f}$, $\tilde{f} > 0$ integer.

Remark 4: To simplify the notation used throughout the upcoming results and to explicitly focus this article on the formulation of a nonlinear moment-based controller, it is assumed that the moment-domain equivalent L_e , characterizing the wave excitation \mathcal{F}_e as in (17), is known, i.e., full (instantaneous and future) knowledge of \mathcal{F}_e is available over the time interval $\Xi \subset \mathbb{R}^+$. This is without loss of generality since estimation and forecasting algorithms for \mathcal{F}_e (which are often required due to the inherent difficulty behind measuring wave excitation forces in a moving body [31]) can be incorporated straightforwardly, by following the adaptation of the moment-based representation of \mathcal{F}_e for receding-horizon control presented in [32, Sec. IV-A], without further modifications.

⁸Similar arguments can be adopted for multi-DoF systems. The reader is referred to [16] and [29] for further detail.

Even though the wave excitation force is composed of \tilde{f} harmonics multiple of the (angular) fundamental frequency ω_0 , it is convenient (for the subsequent theoretical results) to assume that the control input u can be composed of a higher number $\tilde{f} + d$ of harmonics, with d > 0 integer. For this purpose, we define the auxiliary "extended" signal generator as follows. Let $\tilde{S} \in \mathbb{R}^{(v+i) \times (v+i)}$ be such that

$$\bar{S} = S \oplus \left(\bigoplus_{p=1}^{d} \begin{bmatrix} 0 & (p+\tilde{f})\omega_0 \\ -(p+\tilde{f})\omega_0 & 0 \end{bmatrix} \right)$$
(19)

with i = 2d. We can now express the wave excitation force and the control input u as a function of this extended signal generator as

$$\dot{\bar{\xi}} = \bar{S}\bar{\xi}$$

$$\mathcal{F}_e = [L_e \ 0]\bar{\xi} = \bar{L}_e\bar{\xi}$$

$$u = \bar{L}_u\bar{\xi}$$
(20)

where $\bar{\xi}(t) \in \mathbb{R}^{\nu+i}$ and $\bar{\xi}(0) = [\xi(0)^{\mathsf{T}}, \xi^{\star}(0)^{\mathsf{T}}]^{\mathsf{T}}, \xi^{\star}(0) \in \mathbb{R}^{i}$. Note that the input defined in (15) can be expressed accordingly as $\mathscr{U} = (\bar{L}_{e} - \bar{L}_{u})\bar{\xi}$.

Remark 5: The signal generator (20) is an extension of the one defined in (17) in the sense that it inherently incorporates the matrix *S* while adding *d* harmonics multiple of the fundamental frequency ω_0 . With the selected initial condition $\bar{\xi}(0)$, the wave excitation force \mathcal{F}_e can be written as a function of $\bar{\xi}$ by simply using an appropriate inclusion mapping, i.e., completing L_e with zeros accordingly.

In preparation for the upcoming results, we introduce, without loss of generality, three further assumptions. The first one is on the signal generator defined in (20).

Assumption 4: The triple of matrices $(\bar{L}_e - \bar{L}_u, \bar{S}, \bar{\xi}(0))$ is minimal, i.e., observable and excitable.

Note that the previous assumption is without loss of generality as the signal generator is user-defined, and therefore, it can always be constructed such that the assumption holds.

Remark 6: Let $\bar{\xi}_i$ be the *i*th entry of $\bar{\xi}$, with $i \in \mathbb{N}_{\nu+i}$, and define the set $\tilde{\mathscr{X}} = \{\bar{\xi}_i\}_{i=1}^{\nu+i}$. Note that, if Assumption 4 holds, then the pair $(\bar{S}, \bar{\xi}(0))$ is excitable and it is straightforward to check that span $\{\tilde{\mathscr{X}}\}$ = span $\{\{\cos(p\omega_0 t), -\sin(p\omega_0 t)\}_{p=1}^{\bar{f}+d}\}$. As a consequence, the input \mathscr{U} is always *T*-periodic, where $T = 2\pi/\omega_0 \in \mathbb{R}^+$ is the fundamental period of \mathscr{U} .

We now state the following standard assumption on the nonlinear mapping f_{nl} , to later prove existence and uniqueness of the moment of system (14) at the signal generator $(\bar{S}, \bar{L}_e - \bar{L}_u)$.

Assumption 5: The mapping $f_{nl} : \mathbb{R}^2 \to \mathbb{R}^2$ is such that

$$f_{nl}(0) = 0, \qquad \left. \frac{\partial f_{nl}(\varphi)}{\partial \varphi} \right|_{\varphi=0} = 0.$$
 (21)

Note that the assumption is without loss of generality because the matrices in (14) and the mapping f_{nl} can always be redefined to satisfy it. Finally, we introduce an assumption on the stability in the first approximation of system (14).

Assumption 6: The zero equilibrium of system

$$\dot{\varphi} = A\varphi - B(k * C\varphi) \tag{22}$$

is asymptotically stable.

As discussed in several studies, such as [24] and [33], the linear equation of motion (22) is asymptotically stable for any meaningful values of the involved parameters (and impulse response function k). Thus, this assumption is, in practice, also without loss of generality.

Proposition 1: Suppose that Assumptions [4]–[6] hold. Then, there exists a mapping π , which solves the partial differential equation

$$\frac{\partial \pi(\xi)}{\partial \bar{\xi}} \bar{S} \bar{\xi} = f(\pi(\bar{\xi}), (\bar{L}_e - \bar{L}_u)\bar{\xi}).$$
(23)

Moreover, the moment of system (14) at the signal generator $(\bar{S}, \bar{L}_e - \bar{L}_u)$ computed along a particular trajectory $\bar{\xi}(t)$ coincides with the well-defined steady-state output response of the interconnected system (14)–(20), i.e., $y_{ss}(t) = h(\pi(\bar{\xi}(t)))$.

Proof: Note that, under Assumption 4, the triple of matrices $(\bar{L}_e - \bar{L}_u, \bar{S}, \bar{\xi}(0))$ is minimal. Moreover, the extended signal generator defined in (20) is such that $\lambda(S) \subset \mathbb{C}^0$ with simple eigenvalues, in line with Assumption 3. Therefore, Proposition 1 automatically holds as long as the zero equilibrium of the system $\dot{\varphi} = f(\varphi, 0)$ is locally exponentially stable (see Theorem 1). Since this is the case by Assumption 6, the proof is concluded.

In slightly different words, Proposition 1 guarantees that the steady-state response of system (14), driven by (20), can be effectively computed using the corresponding moment at $(\bar{S}, \bar{L}_e - \bar{L}_u)$. Nevertheless, even if the existence of π [solution of (23)] is guaranteed, it is virtually impossible to compute its analytic expression when the mapping f in (14) is nonlinear.

The very nature of the mapping π intrinsically depends on both the characteristics of the signal generator (20) and the system dynamics defined by f. Aiming to formally characterize π , we introduce the following key remarks, which drives the next main result.

Remark 7: Let Ξ be defined as $\Xi = [0, T] \subset \mathbb{R}^+$. Note that the set $\tilde{\mathscr{X}}$, defined in Remark 6, belongs to the Hilbert space $L^2(\Xi)$ and is orthogonal under the standard inner product operator. Moreover, if Assumption 4 holds, we can always complete $\tilde{\mathscr{X}}$ to an orthogonal basis \mathscr{X} of $L^2(\Xi)$, i.e., we define (see [34, Ch. 8])

$$\mathscr{X} = \bar{\mathscr{X}} \cup \check{\mathscr{X}}$$

 $\check{\mathscr{X}} = \{\cos(p\omega_0 t), -\sin(p\omega_0 t)\}_{p=\tilde{f}+d+1}^{\infty} = \{\check{\mathscr{X}}_i\}_{i=\nu+i+1}^{\infty}.$

Remark 8: If Assumption 4 holds, one can always find a set of mappings $\mathcal{I}_i : \mathbb{R}^{\nu+i} \to \mathbb{R}$ such that $\check{\mathscr{X}}_i = \mathcal{I}_i(\bar{\zeta})$, for every $i > \nu + i$ integer. This (standard) result states that we can always generate the elements of the set $\check{\mathscr{X}}$ (i.e., higher order harmonics of the fundamental frequency ω_0) by solely operating on the $\nu + i$ trigonometric polynomials defined by the entries of $\bar{\zeta}$ (see [35]).

Proposition 2: Suppose that Assumptions [4]–[6] hold. Then, each element of the mapping π , as the solution to (23), i.e., π_k , $k \in \mathbb{N}_2$, belongs to the Hilbert space $L^2(\Xi)$ with $\Xi = [0, T] \subset \mathbb{R}^+$, where $T = 2\pi/\omega_0$, i.e., it can be uniquely expressed as

$$\pi_k(\bar{\xi}) = \sum_{i=1}^{\nu+i} \alpha_{k_i} \bar{\xi}_i + \epsilon_k = \bar{\Pi}_k \bar{\xi} + \epsilon_k \tag{24}$$

where $\epsilon_k = \sum_{i=\nu+i+1}^{\infty} \alpha_{k_i} \mathcal{I}_i(\bar{\xi}), \ \alpha_{k_i} \in \mathbb{R} \forall i$, with \mathcal{I}_i as defined in Remark 8, and the matrix $\bar{\Pi}_k^{\mathsf{T}} \in \mathbb{R}^{\nu+i}$ is given by $\bar{\Pi}_k = [\alpha_{k_1}, \ldots, \alpha_{k_{\nu+i}}].$

Proof: Given the nature of the signal generator defined in (20), the function \mathscr{U} is *T*-periodic, with $T = 2\pi/\omega_0$ (see Remark 6). Moreover, under the above assumptions, the zero equilibrium of $\dot{\varphi} = f(\varphi, 0)$ is locally exponentially stable and its (well-defined) steady-state solution is also *T*-periodic [36, Sec. VI], i.e., $\varphi_{ss}(t) = \varphi_{ss}(t - T)$. Since under Assumptions [4]–[6], $\varphi_{ss}(t) = \pi(\bar{\xi}(t))$ (see Proposition 1), it is straightforward to conclude that each element of the mapping π belongs to $L^2(\Xi)$, i.e., it can be expressed as a unique linear combination of the orthogonal basis \mathscr{X} (as defined in Remark 7), which concludes the proof.

Remark 9: The result of Proposition 2 allows π to be compactly expressed as

$$\pi(\bar{\xi}) = \begin{bmatrix} \bar{\Pi}_1 \\ \bar{\Pi}_2 \end{bmatrix} \bar{\xi} + \begin{bmatrix} \epsilon_1 \\ \epsilon_2 \end{bmatrix} = \bar{\Pi}\bar{\xi} + E$$
(25)

where the term $E : \mathbb{R}^{\nu+i} \to \mathbb{R}^2$ is called the truncation error. Note that, if we "ignore" the truncation error E, the mapping π can be effectively approximated as $\bar{\pi}(\bar{\zeta}) = \bar{\Pi}\bar{\zeta}$, i.e., by its expansion on the $(\nu + i)$ -dimensional set $\tilde{\mathscr{X}}$. This motivates the following key definition.

Definition 4: We call the function $C\bar{\pi}$, where $\bar{\pi}(\bar{\xi}) = \bar{\Pi}\bar{\xi}$, the approximated moment⁹ of system (14) at the signal generator $(\bar{S}, \bar{L}_e - \bar{L}_u)$. In addition, we refer to the matrix $\bar{Y} = C\bar{\Pi}$ as the approximated moment-domain equivalent¹⁰ of y.

Remark 10: Under the same set of assumptions as Proposition 2, the approximated moment-domain equivalent of y can be effectively used to approximate the steady-state output of system (14) driven by $(\bar{S}, \bar{L}_e - \bar{L}_u)$, i.e., $y_{ss}(t) \approx C \bar{\Pi} \bar{\xi}(t) = \bar{Y} \bar{\xi}(t)$.

Aiming to propose a method to compute \overline{Y} and inspired by the family of mean weighted residual methods [38], [39], we define the residual mapping $\mathcal{R} : \mathbb{R}^2 \to \mathbb{R}^2$ as

$$\mathcal{R}(\bar{\Pi}\bar{\xi}) := \bar{\Pi}\bar{S}\bar{\xi} - f(\bar{\Pi}\bar{\xi}, (\bar{L}_e - \bar{L}_u)\bar{\xi})$$
(26)

which directly arises from "replacing" π with $\bar{\pi}$ in (23). Using this residual equation, we consider a collocation approach [39, Ch. 4] to compute the approximated moment-domain equivalent $\bar{Y} = C\bar{\Pi}$. In other words, we force (26) to be exactly zero at a finite set of collocation points. We make this approximation method explicit in the following proposition.

Proposition 3: Consider the nonlinear system (14) and the signal generator defined by (20). Suppose that Assumptions [4]–[6] hold. Then, the approximated moment-domain equivalent of y can be computed as $C\Pi$, where Π is the solution of the algebraic system of equations

$$(\bar{\Pi}\bar{S} - A\bar{\Pi} + BC\bar{\Pi}\mathscr{R} - B(\bar{L}_e - \bar{L}_u))\langle\bar{\xi}, \delta_{t_j}\rangle - \langle f_{nl}(\bar{\Pi}\bar{\xi}), \delta_{t_j}\rangle = 0$$
(27)

⁹This notion is analogous to the one given in [37].

¹⁰This definition is analogous to that used in the linear moment-based WEC control studies [15], [16].

with $\mathcal{T}_{\delta} = \{t_i\}_{i=1}^{\nu+\iota} \subset \Xi$ a set of uniformly distributed time instants, and where the matrix $\mathscr{R} \in \mathbb{R}^{(\nu+\iota)\times(\nu+\iota)}$ is defined as

$$\mathscr{R} = \bigoplus_{p=1}^{f+d} \begin{bmatrix} \Re\{K(p\omega_0)\} & \Im\{K(p\omega_0)\}\\ -\Im\{K(p\omega_0)\} & \Re\{K(p\omega_0)\} \end{bmatrix}$$
(28)

with $K : \mathbb{R} \to \mathbb{C}$, $\omega \mapsto K(\omega)$, the Fourier transform of the impulse response function associated with radiation effects, i.e., *k* in (14).

Proof: Note that, using (14), the residual equation (26) can be equivalently written as

$$(\bar{\Pi}\bar{S} - A\bar{\Pi} - B(\bar{L}_e - \bar{L}_u))\bar{\xi} + B(k * C\bar{\Pi}\bar{\xi}) - f_{nl}(\bar{\Pi}\bar{\xi})$$
(29)

where the convolution operation involved, associated with the effect of radiation forces acting on the device, can be shown to be such that [15], [16]

$$k * C \bar{\Pi} \bar{\xi} = C \bar{\Pi} \mathscr{R} \bar{\xi} \tag{30}$$

with \mathscr{R} as in (28). Then, following the well-known collocation method [39, Ch. 4], the residual function is forced to be orthogonal (under the standard inner-product of $L^2(\Xi)$) to the set of translated Dirac- δ functions $\{\delta_{t_i}\}_{i=1}^{\nu+i}$. Equation (27) follows after considering the superposition property of the inner product operator.

Corollary 1: The system of algebraic equations (27) can be equivalently written in the matrix form as

$$(\bar{\Pi}\bar{S} - A\bar{\Pi} + BC\bar{\Pi}\mathcal{R} - B(\bar{L}_e - \bar{L}_u)) - F_{nl}(\bar{\Pi})\Omega^{-1} = 0$$
(31)

where the matrices $F_{nl}(\bar{\Pi}) \in \mathbb{R}^{2 \times (\nu+i)}$ and $\Omega \in \mathbb{R}^{(\nu+i) \times (\nu+i)}$ are defined as

$$\Omega = \begin{bmatrix} \bar{\xi}(t_1) & \cdots & \bar{\xi}(t_{\nu+l}) \end{bmatrix}$$

$$F_{nl}(\bar{\Pi}) = \begin{bmatrix} f_{nl}(\bar{\Pi}\bar{\xi}(t_1)) & \cdots & f_{nl}(\bar{\Pi}\bar{\xi}(t_{\nu+l})) \end{bmatrix}. \quad (32)$$

Proof: Note that if the set $\{t_j\} \subset \Xi$, then $\langle f, \delta_{t_j} \rangle = f(t_j)$, for any continuous function $f : \Xi \to \mathbb{R}$. Then, the result follows as a consequence of the excitability of the pair $(\bar{S}, \xi(\bar{0}))$, which implies that the matrix Ω is always full rank [23].

Remark 11: If the set of uniformly distributed time instants \mathcal{T}_{δ} is chosen such that $t_k = -T/2 + Tk/(\nu + \iota)$, $t_k \in \mathcal{T}_{\delta}$ for all $k \in \mathbb{N}_{\nu+\iota}$, then the collocation approach utilized in Proposition 3 is identical to the Galerkin method [39, Ch. 4]. The main advantage of Proposition 3 (collocation) lies in its simplicity of implementation, i.e., we simply use function evaluation (see Corollary 1).

Remark 12: In the light of Remark 11, standard results of Galerkin methods (see [40]) apply to this WEC case. In particular, the existence of solutions to system (27) [equivalently (31)], under the hypothesis of Proposition 3, is always guaranteed for all sufficiently large *i*. Moreover, the approximated moment $\bar{\pi}(\bar{\xi}) = \bar{\Pi}\bar{\xi}$ converges uniformly toward the exact solution (25) as $i \to \infty$ (see also [40]).

We now present a corollary, which illustrates the result of Proposition 3 (through Corollary 1) in a more convenient form for the upcoming nonlinear moment-based energy-maximizing control formulation. In particular, we show that (31) can be fully expressed in terms of the approximated moment-domain equivalent $\bar{Y} = C\bar{\Pi}$, effectively reducing the number of variables involved in such an equation.

Corollary 2: The system of algebraic equations (31) can be fully written as a function of the approximated moment-domain equivalent $\bar{Y} = C\bar{\Pi}$ as

$$\bar{Y} - (\bar{L}_e - \bar{L}_u)\bar{\Phi}^{\mathsf{T}} + \operatorname{vec}\{F_{nl}(g(\bar{Y}))\}^{\mathsf{T}}\Phi_{\Omega}^{\mathsf{T}} = 0$$
(33)

where $\bar{\Phi} \in \mathbb{R}^{(\nu+\iota) \times (\nu+\iota)}$ and $\Phi_{\Omega} \in \mathbb{R}^{(\nu+\iota) \times 2(\nu+\iota)}$ are given by the expressions

$$\bar{\Phi} = (\mathbb{I}_{\nu+\iota} \otimes C) \Phi^{-1} (\mathbb{I}_{\nu+\iota} \otimes -B)
\Phi_{\Omega} = (\mathbb{I}_{\nu+\iota} \otimes C) \Phi^{-1} (\Omega^{-1^{\mathsf{T}}} \otimes \mathbb{I}_{2})
\Phi = \bar{S} \widehat{\oplus} A + \mathscr{R}^{\mathsf{T}} \otimes -BC$$
(34)

and the mapping $g: \mathbb{R}^{1 \times (\nu + i)} \to \mathbb{R}^{2 \times (\nu + i)}$ is defined as

$$g(\bar{Y}) = (\mathbb{I}_2 \otimes \bar{Y}) \begin{bmatrix} \bar{S}^{-1} \\ \mathbb{I}_{\nu+i} \end{bmatrix}.$$
 (35)

Proof: A direct application of the vec operator (and Property 1) to (32) yields

$$\operatorname{vec}\{\bar{\Pi}\} + \Phi^{-1}(\mathbb{I}_{\nu+\iota} \otimes B)\operatorname{vec}\{\bar{L}_e - \bar{L}_u\} + \Phi^{-1}(\Omega^{-1^{\mathsf{T}}} \otimes \mathbb{I}_2)\operatorname{vec}\{F_{nl}(\bar{\Pi})\} = 0 \quad (36)$$

in which we make explicit use of the skew symmetricity of \bar{S} , i.e., $-\bar{S}^{\mathsf{T}} = \bar{S}$ to obtain Φ as in (34). The invertibility of the matrix Φ has been shown in [15] and [16]. Equation (33) follows after multiplying both sides of (36) by $(\mathbb{I}_{\nu+\iota} \otimes C)$, where we note that $\operatorname{vec}\{C\bar{\Pi}\} = \operatorname{vec}\{\bar{Y}\} = \bar{Y}^{\mathsf{T}}$ and $\operatorname{vec}\{\bar{L}_e - \bar{L}_u\} = (\bar{L}_e - \bar{L}_u)^{\mathsf{T}}$. Finally, the mapping g arises as a result of [41, Proposition 1]: given that $\dot{\varphi}_1 = \varphi_2 = y$ in (14), $\bar{\Pi}$ can be written in terms of \bar{Y} simply as

$$\bar{\Pi} = \begin{bmatrix} \bar{Y}\bar{S}^{-1}\\ \bar{Y} \end{bmatrix} = (\mathbb{I}_2 \otimes \bar{Y}) \begin{bmatrix} \bar{S}^{-1}\\ \mathbb{I}_{\nu+i} \end{bmatrix} = g(\bar{Y})$$
(37)

which concludes our proof.

Remark 13: If $F_{nl}(g(\bar{Y})) = 0$, i.e., system (14) is linear, the approach of Proposition 3 (through Corollary 2) recovers (without approximation) the linear moment-domain equivalent $\bar{Y} = (\bar{L}_e - \bar{L}_u)\bar{\Phi}^{\mathsf{T}}$, presented in [15] and [16]. In other words, (33) can be regarded as the linear moment domain equivalent of y plus a nonlinear "perturbation" term.

B. Motion Unconstrained Nonlinear Moment-Based OCP

The results presented in Section IV-A can be effectively used to approximate the energy-maximizing OCP presented in Definition 1, making explicit use of the connection between moments and the steady-state behavior of system (14). In the following, we provide a definition of the so-called approximated energy-maximizing OCP, using the approximated moment-domain equivalent \bar{Y} , presented in Definition 4. Note that we do not yet include the set of state and input constraints defined in (12). These are explicitly incorporated in Section IV-C.

Problem 2 (Approximated Energy-Maximizing OCP): Suppose that Assumptions [4]–[6] hold. Find the optimal control input $\bar{u}^{\text{opt}} = \bar{L}_{u}^{\text{opt}} \bar{\zeta}$ such that

$$\bar{L}_{u}^{\text{opt}} = \arg \max_{\bar{L}_{u}^{\mathsf{T}} \in \mathbb{R}^{\nu+i}} \frac{1}{T} \int_{\Xi} \bar{L}_{u} \bar{\xi}(\tau) \bar{Y} \bar{\xi}(\tau) d\tau$$
s.t.: $\bar{Y} - (\bar{L}_{e} - \bar{L}_{u}) \bar{\Phi}^{\mathsf{T}} + \text{vec}\{F_{nl}(g(\bar{Y}))\}^{\mathsf{T}} \Phi_{\Omega}^{\mathsf{T}} = 0$
(38)

where \overline{Y} is the approximated moment domain equivalent of the output of system (14) (see Definition 4) and $\overline{\xi}$ is the solution of (20).

Remark 14: The main idea behind Problem 2 relies on substituting the integrodifferential (equality) constraint, corresponding with the nonlinear WEC dynamics (14), by the algebraic equation (33). Note that the latter characterizes the approximated moment-domain equivalent of the velocity of the device $\dot{x} = y$ [which is the key state variable involved in the energy-maximizing objective function (8)]. In other words, the approximated OCP posed in Problem 2 explicitly utilizes an approximation of the steady-state (output) behavior of system (14), parameterized in terms of \bar{Y} , i.e., $\dot{x}_{ss} = y_{ss}(t) \approx \bar{Y}\bar{\xi}(t)$ (see Remark 10), to solve for the corresponding optimal control input \bar{u}^{opt} , in terms of the signal generator (20).

Remark 15: Following Remark 12, if $\iota \to \infty$, then the steady-state output response of system (14) is exactly given by $y_{ss}(t) = \bar{Y}\bar{\xi}(t)$ and the algebraic equality constraint in the OCP of Problem 2 corresponds to the exact steady-state motion of the device, without approximations.

Based on Problem 2, we are now ready to propose a solution to the motion unconstrained energy-maximizing OCP, i.e., (13) without considering input and state constraints (see Problem 2), in terms of a specific tractable finite-dimensional NP. This claim is formalized in the following proposition.

Proposition 4 (Nonlinear Moment-Based Unconstrained NP): Suppose that Assumptions [4]–[6] hold, and let $\bar{\xi}(0) = \varepsilon_{\nu+\iota}$. Then, for ι sufficiently large, the solution of the (motion unconstrained) approximated energy-maximizing OCP, posed in Problem 2, can be computed as $\bar{u}^{\text{opt}} = \bar{L}_{u}^{\text{opt}}\xi$, where

$$\bar{L}_{u}^{\text{opt}} = -\left(\bar{Y}^{\text{opt}} + \text{vec}\{F_{nl}(g(\bar{Y}^{\text{opt}}))\}^{\mathsf{T}}\Phi_{\Omega}^{\mathsf{T}} - \bar{L}_{e}\right)\bar{\Phi}^{-1\mathsf{T}}$$
(39)

and the matrix \bar{Y}^{opt} is the solution of the finite-dimensional NP

$$\bar{Y}^{\text{opt}} = \arg \max_{\bar{Y}^{\mathsf{T}} \in \mathbb{R}^{\nu+i}} \bar{\mathcal{J}}_{\text{QP}}(\bar{Y}) + \bar{\mathcal{J}}_{nl}(\bar{Y}) \tag{40}$$

with $\bar{\mathcal{J}}_{QP}: \mathbb{R}^{\nu+l} \to \mathbb{R}, \ \bar{\mathcal{J}}_{nl}: \mathbb{R}^{\nu+l} \to \mathbb{R}$ defined as

$$\bar{\mathcal{J}}_{QP}(\bar{Y}) = -\frac{1}{2}\bar{Y}\bar{\Phi}^{-1}\bar{Y}^{\mathsf{T}} + \frac{1}{2}\bar{Y}\bar{L}_{e}^{\mathsf{T}}
\bar{\mathcal{J}}_{nl}(\bar{Y}) = -\frac{1}{2}\bar{Y}\bar{\Phi}^{-1}\Phi_{\Omega}\operatorname{vec}\{F_{nl}(g(\bar{Y}))\}.$$
(41)

Proof: The fundamental step toward this proof lies in [15, Proposition 3]. In particular, due to the (harmonic) nature of the signal generator defined in (20), the objective function corresponding with the approximated OCP, i.e., (38), can be equivalently written [15] as

$$\bar{\mathcal{J}} = \frac{1}{T} \int_{\Xi} \bar{Y} \bar{\xi}(\tau) \bar{L}_u \bar{\xi}(\tau) d\tau = \frac{1}{2} \bar{Y} \bar{L}_u^{\mathsf{T}}$$
(42)

for $\bar{\xi}(0) = \varepsilon_{\nu+\iota}$. Substituting \bar{L}_u in (42), using the result of Corollary 2, we can write $\bar{\mathcal{J}}$, as a function of \bar{Y} , as

$$\bar{\mathcal{J}} = -\frac{1}{2}\bar{Y}\bar{\Phi}^{-1}\bar{Y}^{\mathsf{T}} + \frac{1}{2}\bar{Y}\bar{L}_{e}^{\mathsf{T}} - \frac{1}{2}\bar{Y}\bar{\Phi}^{-1}\Phi_{\Omega}\operatorname{vec}\{F_{nl}(g(\bar{Y}))\}$$

$$\bar{\mathcal{J}} = \bar{\mathcal{J}}_{\mathrm{QP}}(\bar{Y}) + \bar{\mathcal{J}}_{nl}(\bar{Y})$$
(43)

where the optimal control input $\bar{u}^{\text{opt}} = \bar{L}_{u}^{\text{opt}}\bar{\xi}$ can be straightforwardly recovered using the equality (33), yielding (39), which concludes our proof.

Proposition 4 explicitly uses the approximated moment domain equivalent \bar{Y} to propose a finite-dimensional tractable optimization problem, allowing for the computation of an energy-maximizing control solution for the approximated OCP posed in Problem 2 when the WEC is subject to nonlinear dynamics. Note that there is (almost) no restriction on the nature of the mapping f_{nl} , so that a general class of nonlinear effects can be considered, including complex hydrodynamic nonlinearities, such as those discussed in Section V.

Remark 16: The moment-based NP stated in Proposition 4 has to be carried out over the approximated moment domain equivalent $\bar{Y}^{\mathsf{T}} \in \mathbb{R}^{\nu+l}$ only, i.e., in terms of the moment-domain representation of the velocity of the device, and can be solved using efficient state-of-the-art numerical routines, such as interior-point methods (IPMs) [42].

Remark 17: There is an intrinsic tradeoff between the degree of accuracy behind the approximated OCP, controlled by the parameter ι (see Remark 15), and the underlying computational complexity of (40). In other words, a higher ι results in improved energy absorption but also intrinsically increases the computational requirements of the strategy.

Remark 18: If $\overline{\mathcal{J}}_{nl}(\overline{Y}) = 0$, Proposition 4 recovers the optimal moment-based control input proposed in [15] for the linear WEC case. To be precise, if there are no nonlinearities involved in (14), (40) is of a concave quadratic type, i.e., a QP written as

$$\bar{Y}_l^{\text{opt}} = \arg \max_{\bar{Y}^{\mathsf{T}} \in \mathbb{R}^{\nu+i}} -\frac{1}{2} \bar{Y} \bar{\Phi}^{-1} \bar{Y}^{\mathsf{T}} + \frac{1}{2} \bar{Y} \bar{L}_e^{\mathsf{T}}$$
(44)

where the function $\overline{\mathcal{J}}_{QP}$ is strictly concave for any physically meaningful parameters involved in the WEC equation of motion (14) (see [15, Proposition 4]).

Following Remark 18, it is straightforward to note that the NP stated in Proposition 4 can be seen as a QP problem characterized by the objective function $\overline{\mathcal{J}}_{QP}$ and "perturbed" by the action of the nonlinear mapping $\overline{\mathcal{J}}_{nl}$. Nevertheless, unlike the linear moment-based energy-maximizing OCP of [15] and [16] (recalled herein in Remark 18), there is no guarantee that the nonlinear OCP of Proposition 4 admits a global maximizer. Aiming to secure the existence of a global solution to problem (40), we introduce the following standing assumption, to later formalize an appropriate proposition guaranteeing the existence of a global energy-maximizing solution to (40).

Assumption 7: The mapping $\overline{\mathcal{J}}_{nl}$: $\mathbb{R}^{1 \times (\nu+i)} \to \mathbb{R}$ is bounded by a parameter $\alpha_{nl} \in \mathbb{R}^+$, i.e.,

 \overline{Y}

$$\sup_{\mathbf{r}\in\mathbb{R}^{\nu+i}}|\bar{\mathcal{J}}_{nl}(\bar{Y})|\leq \alpha_{nl}<+\infty.$$
(45)

As demonstrated in Section V, Assumption 7 is without loss of generality since the mapping $\varphi \mapsto \mathcal{F}^{nl}(\varphi)$, which effectively defines the nonlinear nature of $\overline{\mathcal{J}}_{nl}$ [see (41)], is sufficiently smooth (see Section III-A) and hence bounded on any compact set.

Proposition 5: Suppose that Assumption 7 holds. Then, the NP with objective function $\overline{\mathcal{J}}$ defined in Proposition 4 always admits a global maximum $\overline{Y}^{\text{opt}}$.

Proof: The key concept behind this proposition lies in the decomposition of $\overline{\mathcal{J}}$ as in (41), i.e., as the sum of a concave problem $\overline{\mathcal{J}}_{QP}$ and, under Assumption 7, a bounded perturbation $\overline{\mathcal{J}}_{nl}$. To be precise, if $\overline{\mathcal{J}}_{nl}$ is bounded, then the function $-\overline{\mathcal{J}}$ is strictly outer Γ -convex (see Definition 5 and Theorem 2 in the Appendix), for $\Gamma \subset \mathbb{R}^{1 \times (\nu + i)}$, where the set Γ is given by

$$\Gamma = \mathscr{B}(0, r), \quad r = \sqrt{\frac{2\alpha_{nl}}{\min \lambda(\mathscr{H}\{\bar{\Phi}^{-1}\})}}.$$
 (46)

Finally, given that the mapping $\overline{\mathcal{J}}_{QP}$ has a unique global maximizer $\overline{Y}_l^{\text{opt}}$ (see Remark 18) and the set $\Gamma \cap \mathbb{R}^{1 \times (\nu+i)}$ is closed, the NP defined by the objective function $\overline{\mathcal{J}}$ always admits a global optimal solution $\overline{Y}^{\text{opt}}$ (see [43, Lemma 4.3] and Property 3 in the Appendix).

Proposition 5 makes explicit use of the strictly outer Γ convexity of the function $-\overline{\mathcal{J}}$ to ensure the existence of a global solution to the moment-based energy-maximizing OCP proposed in this study. In other words, in contrast to state-of-the-art nonlinear WEC control techniques [9]–[13], we explicitly guarantee the existence of globally optimal solutions to the associated energy-maximizing OCP.

Remark 19: Informally, the concept of outer Γ -convexity, introduced in [19], is a formalization of the concept of convexity for a class of functions that are "roughly" convex, i.e., they possess properties similar to those of convex functions [43]. The reader is referred to Definition 5 in the Appendix for a formal treatment of this concept.

Moreover, recalling key theoretical results from [43], we can use the following property of strictly outer convex functions (see Property 4 in the Appendix), which establishes a direct relationship between local and global maximizers for $\overline{\mathcal{J}}$, having strong practical implications.

Property 2 [43]: Let $\overline{Y}^{\text{opt}}$ be a Γ-local maximizer of $\overline{\mathcal{J}}$, i.e.,

$$\bar{\mathcal{J}}(\bar{Y}^{\text{opt}}) = \max_{\bar{Y} \in \mathscr{B}(\bar{Y}^{\text{opt}}, r)} \bar{\mathcal{J}}(\bar{Y})$$
(47)

with r as in (46). Then, \bar{Y}^{opt} is a global maximizer of $\bar{\mathcal{J}}$.

Property 2 (or, analogously, Property 4 in the Appendix) acts as the analog of the global optimality property of concave functions (i.e., every local solution is a global solution). In other words, if \bar{Y}^{opt} is a maximizer for $\mathscr{B}(\bar{Y}^{\text{opt}}, r)$, a subset of $\mathbb{R}^{1 \times (\nu+i)}$, then it is automatically a global maximizer of $\bar{\mathcal{J}}$. This not only gives explicit conditions for global energy maximization within our nonlinear moment-based approach (in contrast to available nonlinear WEC optimal control techniques) but also considerably reduces the "search" space when numerically solving (40), enhancing the efficiency behind the proposed moment-based strategy.

C. On the Inclusion of State and Input Constraints

As discussed in Section III-B, any energy-maximizing optimal control strategy must consider physical limitations, arising from both the device itself, and the actuator (PTO system) dynamics. Following the moment-based NP defined in Proposition 4, we propose a framework to incorporate the set of state and input constraints (12) to the energy-maximizing unconstrained solution of Proposition 4.

To be precise, and in line with [15] and [16], we map the set of constraints defined in (12) onto their respective moment domain equivalents as

$$\begin{cases} |x(t)| \leq X_{\max} \\ |\dot{x}(t)| \leq V_{\max} \\ |u(t)| \leq U_{\max} \end{cases} \mapsto \begin{cases} |\bar{Y}S^{-1}\bar{\xi}(t)| \leq X_{\max} \\ |\bar{Y}\bar{\xi}(t)| \leq V_{\max} \\ |\bar{L}_{u}\bar{\xi}(t)| \leq U_{\max} \end{cases}$$
(48)

and we enforce them only at a finite set of N_{ρ} uniformly spaced time instants $\mathcal{T}_{\rho} = \{t_i\}_{t=1}^{N_{\rho}} \subset \Xi$, i.e., using a collocation approach. To that end, we define the matrices $\bar{\Lambda} \in \mathbb{R}^{(\nu+i) \times N_{\rho}}$ and $\bar{\Delta} \in \mathbb{R}^{(\nu+i) \times 2N_{\rho}}$ as

$$\bar{\Lambda} = \begin{bmatrix} \bar{\xi}(t_1) & \cdots & \bar{\xi}(t_{N_{\rho}}) \end{bmatrix}, \quad \bar{\Delta} = \begin{bmatrix} \bar{\Lambda} & -\bar{\Lambda} \end{bmatrix}.$$
(49)

Finally, we can formulate a moment-based energy-maximizing constrained optimal control solution for WECs, subject to nonlinear dynamics, as follows.

Corollary 3 (Nonlinear Moment-Based Constrained NP): Suppose that Assumptions [4]–[6] hold, and let $\bar{\xi}(0) = \varepsilon_{\nu+i}$. Then, for *i* sufficiently large, the solution of the approximated energy-maximizing OCP, posed in Problem 2, subject to the set of state and input constraints (48), can be computed as $\bar{u}^{\text{opt}} = \bar{L}_{u}^{\text{opt}}\xi$, where

$$\bar{L}_u^{\text{opt}} = -\left(\bar{Y}^{\text{opt}} + \text{vec}\{F_{nl}(g(\bar{Y}^{\text{opt}}))\}^{\mathsf{T}}\Phi_{\Omega}^{\mathsf{T}} - \bar{L}_e\right)\bar{\Phi}^{-1^{\mathsf{T}}}$$
(50)

and the matrix \bar{Y}^{opt} is the solution of the inequality-constrained finite-dimensional NP

$$\bar{Y}^{\text{opt}} = \arg \max_{\bar{Y}^{\mathsf{T}} \in \mathbb{R}^{\nu+i}} \bar{\mathcal{J}}_{\text{QP}}(\bar{Y}) + \bar{\mathcal{J}}_{nl}(\bar{Y})$$
(51)

s.t.:
$$\begin{cases} Y\mathcal{A}_{x} \leq \mathcal{B}_{x} \\ \bar{Y}\mathcal{A}_{\dot{x}} \leq \mathcal{B}_{\dot{x}} \\ \bar{Y}\mathcal{A}_{u} + \mathcal{N}_{u}(\bar{Y}) \leq \mathcal{B}_{u} \end{cases}$$
(52)

where

$$\mathcal{A}_{x} = \bar{S}^{-1}\bar{\Delta}, \qquad \mathcal{B}_{x} = X_{\max}\mathbf{1}_{1\times 2N_{\rho}}$$
$$\mathcal{A}_{\dot{x}} = \bar{\Delta}, \qquad \mathcal{B}_{\dot{x}} = V_{\max}\mathbf{1}_{1\times 2N_{\rho}}$$
$$\mathcal{A}_{u} = -\bar{\Phi}^{-1^{\mathsf{T}}}\bar{\Delta}, \qquad \mathcal{B}_{u} = U_{\max}\mathbf{1}_{1\times 2N_{\rho}} + \bar{L}_{e}\mathcal{A}_{u}$$
$$\mathcal{N}_{u}(\bar{Y}) = -\operatorname{vec}\{F_{nl}(g(\bar{Y}))\}^{\mathsf{T}}\Phi_{\Omega}^{\mathsf{T}}\mathcal{A}_{u}. \qquad (53)$$

Proof: Note that under the set of assumptions considered in this corollary, (50) and (51) follow directly from Proposition 4. With respect to the incorporation of the set of state and input constraints defined in (48), let us first consider the constraint associated with the control input, and note that

$$|\bar{L}_{u}S^{-1}\bar{\xi}(t)| \le U_{\max} \Rightarrow -U_{\max} \le \bar{L}_{u}\bar{\xi}(t) \le U_{\max}.$$
 (54)

Equation (54), enforced at the set of collocation instants T_{ρ} , can be straightforwardly written in terms of the matrix $\overline{\Delta}$ defined in (49), i.e.,

$$\bar{L}_u \bar{\Delta} \le U_{\max} \mathbf{1}_{1 \times 2N_{\theta}}.$$
(55)

The left-hand side of (55) can be expanded using the result of Corollary 2 as

$$\bar{L}_{u}\bar{\Delta} = \bar{Y}\mathcal{A}_{u} - \bar{L}_{e}\mathcal{A}_{u} + \operatorname{vec}\{F_{nl}(g(\bar{Y}))\}^{\mathsf{T}}\Phi_{\Omega}^{\mathsf{T}}\mathcal{A}_{u} \qquad (56)$$

from where both the matrix \mathcal{B}_u and the nonlinear mapping \mathcal{N}_u follow directly. Finally, the claim of this corollary follows by writing the set of constraints associated with displacement and velocity, defined in (48), as in (55), i.e., in terms of the matrix $\overline{\Delta}$.

Remark 20: The set of inequality constraints associated with the displacement and the velocity is linear in \bar{Y} . This is not the case for the control input-related constraint, which can be decomposed as the sum of a linear and a nonlinear mapping $\mathcal{N}_{\mu} : \mathbb{R}^{1 \times (\nu+i)} \to \mathbb{R}^{1 \times 2N_{\rho}}$.

Remark 21: Other types of constraints different from those considered in (48), e.g., limits on maximum rated power [44], unidirectional power flow [11], or specific nonlinear constraints on the control force arising from selecting a specific PTO system [45], can be incorporated into the presented moment-based approach by following an analogous procedure to that described in this section. In other words, these constraints can be included by a suitable mapping onto their respective moment-domain equivalents, followed by a (time) collocation approach.

V. CASE STUDY: A CorPower-LIKE DEVICE

To demonstrate the performance of the nonlinear momentbased controller proposed in Section IV-B, we consider a full-scale state-of-the-art CorPower-like wave energy device oscillating in heave (translational motion). The actual Cor-Power device is a wave energy system currently under development by the Swedish company CorPower Ocean¹¹, with the aim of making a mass and volume-efficient solution using bottom-referenced heaving buoys. The development of such a WEC builds heavily on research results and earlier experience, and the interested reader is referred to, for instance, [47] and [48]. Note that this type of device is often considered as a case study, due to its intrinsic geometrical complexity (see [49]). Fig. 2 shows a schematic of the CorPower-like WEC, along with its corresponding hydrodynamic characterization, i.e., the frequency-response $K(j\omega)$ associated with the impulse response mapping k. Note that $K(j\omega)$ has been computed using the boundary element method solver NEMOH [50]. The dimensions of this device are based on the experimental study performed in [48].

In the remainder of this section, we consider waves generated stochastically from a JONSWAP spectrum [51], with fixed significant wave height H_s of 2 [m], varying peak period $T_p \in [5, 12]$ [s], peak shape parameter $\gamma = 3.3$, and a total time length (fundamental period) of T = 120 [s]. The

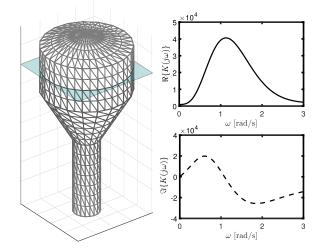


Fig. 2. Schematic of the CorPower-like device, along with the frequency response of the radiation impulse response mapping k. The still water level is indicated using a blue-colored (shaded) plane.

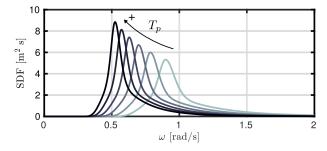


Fig. 3. SDFs for different JONSWAP spectra with fixed wave height of $H_s = 2$ [m] and varying peak period $T_p \in [5, 12]$ (the direction of increasing T_p is indicated by the arrow). The peak shape parameter is fixed to $\gamma = 3.3$.

corresponding spectral density functions (SDFs) are shown, for reference, in Fig. 3. Since the waves are generated from sets of random amplitudes [30], it is found that a mean of ≈ 40 simulations (per sea state) is necessary to obtain statistically consistent performance results for the nonlinear moment-based controller presented in this study, particularly those discussed in Fig. 5.

A. Characterization of Nonlinear Hydrodynamic Effects

In this section, we characterize the nonlinear effects associated with the CorPower-like device, i.e., the mapping \mathcal{F}^{nl} in (11), considered for this case study. In particular, we consider two main hydrodynamic forces: viscous effects \mathcal{F}_v and the presence of a nonlinear restoring force \mathcal{F}_r^{nl} , so that $\mathcal{F}^{nl} = \mathcal{F}_v + \mathcal{F}_r^{nl}$. We give explicit motivation behind the consideration of these nonlinear effects in the upcoming paragraphs.

Viscous effects, arising from vortex shedding and turbulence, are particularly present in heaving point absorber devices [52], such as the CorPower-like WEC considered. One common way of including viscous drag effects in the equation of motion is by means of a Morison-like term [53], i.e., we define

$$\mathcal{F}_{v}(\dot{x}) = -\beta_{v} |\dot{x}| \dot{x} \tag{57}$$

where $\beta_v = (1/2)\rho\gamma_d D$, $\gamma_d \in \mathbb{R}^+$ is the so-called drag coefficient, and D is the characteristic area of the device (as defined in Section III-A). The drag coefficient is set to

¹¹See [46] for up-to-date information on the CorPower device.

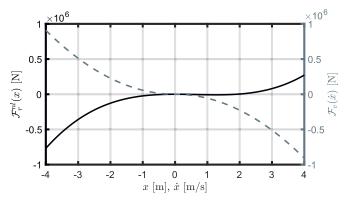


Fig. 4. Nonlinear hydrodynamic effects considered in this study: hydrostatic force (displacement-dependent, left axis), and viscous force (velocitydependent, right axis).

 $\gamma_d = 0.35$, based on the analysis performed in [7] for the device of Fig. 2. Fig. 4 (right axis, dashed line) shows the output of the mapping \mathcal{F}_v .

The mapping \mathcal{F}_r^{nl} , characterizing nonlinear restoring effects, is computed based on the experimental results presented in [48] for this full-scale CorPower device. In particular, inspired by [48], we define

$$\mathcal{F}_{r}^{nl}(x) = \beta_{r_1} x^2 + \beta_{r_2} x^3 \tag{58}$$

where the coefficients $\{\beta_{r_1}, \beta_{r_2}\} \subset \mathbb{R}$ are determined based on a least-squares fit, using the experimental results given in [48] as target set, giving a final result of $\beta_{r_1} = -1.55 \times 10^4$ [kg/ms²] and $\beta_{r_2} = 0.82 \times 10^4$ [kg/m²s²]. The output of the nonlinear restoring force mapping \mathcal{F}_r^{nl} is presented in Fig. 4 (left axis, solid line).

Note that both nonlinear effects, as described in (57) and (58), fulfill Assumption 5. To show that Assumption 7 holds, recall that the energy-maximizing optimal control law is such that the state-variables $\varphi_1 = x$ and $\varphi_2 = \dot{x}$ have maximum allowed values X_{max} and V_{max} , respectively [see (48)]. Then, the following inequality, involving the nonlinear mapping $\mathcal{F}^{nl} = \mathcal{F}_v + \mathcal{F}_r^{nl}$

$$\begin{aligned} |\mathcal{F}^{nl}(\varphi)| &= \beta_v |\varphi_2|^2 + \beta_1 |\varphi_1|^2 + \beta_2 |\varphi_2|^3 \\ &\leq \beta_v V_{\max} + (\beta_1 + \beta_2 X_{\max}) X_{\max}^2 = \tilde{\alpha}_{nl} \end{aligned} (59)$$

holds for all $t \in \Xi$. Using (59), and considering well-known (Euclidean) norm properties, it is straightforward to show that

$$\|\operatorname{vec}\{F_{nl}(g(\bar{Y}))\}\|_{2} \le (\nu + \iota)\tilde{a}_{nl}^{2}.$$
 (60)

Recalling from the set of moment-domain constraints (48) that $|\bar{Y}\bar{\xi}(t)| \leq V_{\max}$ for all $t \in \Xi$, and, if $\bar{\xi}(0) = \varepsilon_{\nu+i}$, then $\|\bar{\xi}\|_2^2 = \sum_{i=1}^{\nu+i} \langle \bar{\xi}_i, \bar{\xi}_i \rangle = T(\nu+i)/2$, we can directly obtain the following estimate for α_{nl} in (45):

$$\begin{aligned} |\tilde{\mathcal{J}}_{nl}(\bar{Y})| &\leq \frac{1}{2T} (V_{\max} \| \bar{\Phi}^{-1} \Phi_{\Omega} \|_F \tilde{\alpha}_{nl})^2 \\ &= \alpha_{nl} < +\infty \end{aligned}$$
(61)

and, hence, the moment-based energy-maximizing OCP always admits a global maximizer under the effect of the nonlinear dynamics defined in \mathcal{F}^{nl} (see Proposition 5).

B. Results and Discussion

Based on the CorPower-like device of Fig. 2, subject to the nonlinear effects described in Section V-A, we now present and discuss the results of applying the nonlinear momentbased energy-maximizing control strategy developed throughout Section IV, under the effect of irregular (polychromatic) wave excitation forces.

We begin by setting the maximum allowed displacement and velocity values as $X_{\text{max}} = 2$ [m] and $V_{\text{max}} = 2$ [m/s]. The wave excitation force \mathcal{F}_e is computed using $\nu = 60$ components in (18), while the order of the extended signal generator (20) is set to v + i = 100. Note that the latter effectively defines the "size" of the optimization variable, i.e., $\bar{Y}^{\mathsf{T}} \in \mathbb{R}^{\nu+i}$. With respect to the collocation instants used to enforce the constraints, as in Section IV-C, we have selected ten uniformly distributed collocation points per second of simulation, i.e., $N_{\rho} = 1200$. The constrained moment-based OCP stated in Corollary 3 can be solved using a variety of state-of-the-art numerical routines, belonging to the families of both local and global optimization methods.

In this study, we opt for a local IPM, where we take explicit advantage of the strict outer convexity of the energy-related objective function when mapped to the moment-domain; we use Property 2 to numerically ensure that the (potentially local) solution computed with IPMs is, effectively, a global energy maximizer. In particular, we propose the following simple algorithm, written in pseudocode.

1: init algorithm

- 2: Set initial guess $\bar{Y}_0^{\text{opt}} = \bar{Y}_l^{\text{opt}}$;
- 3: global = 0:
- 4: while global $\neq 1$ do
- $\bar{Y}^{\text{opt}} \leftarrow$ Solve the OCP (51) using IPM with \bar{Y}_0^{opt} ; 5:
- Generate a random set $\mathcal{P} = \{\bar{Y}_i\}_{i=1}^P$ with P > 06: (integer) elements, such that $\mathcal{P} \subset \mathscr{B}(\bar{Y}^{\text{opt}}, r)$ and where \overline{Y}_i is such that (52) holds for all $i \in \mathbb{N}_P$;
- if $\overline{\mathcal{J}}(\overline{Y}^{\text{opt}}) \leq \overline{\mathcal{J}}(\overline{Y}_i)$ for all $\overline{Y}_i \in \mathcal{P}$ then 7:
- global = 1;8:
- else if $\exists \underline{Y}_i \in \mathcal{P}$ such that $\overline{\mathcal{J}}(\overline{Y}_i) \leq \overline{\mathcal{J}}(\overline{Y}^{\text{opt}})$ then $\overline{Y}_0^{\text{opt}} = \overline{Y}_i$; 9:
- 10:
- 11: end if
- 12: end while

13:
$$\bar{L}_{u}^{\text{opt}} = -(\bar{Y}^{\text{opt}} + \text{vec}\{F_{nl}(g(\bar{Y}^{\text{opt}}))\}^{\mathsf{T}}\Phi_{\Omega}^{\mathsf{T}} - \bar{L}_{e})\bar{\Phi}^{-1^{\mathsf{T}}};$$

14: $\bar{u}^{\text{opt}} = L_u^{\text{opt}} \xi;$ 15: end algorithm.

Starting from the linear solution \bar{Y}_{l}^{opt} of the concave QP problem of Corollary 3, i.e., with $F_{nl}(g(\bar{Y})) = 0$, this heuristic attempts to compute a local solution using IPMs and simply uses function evaluation at a finite set of P random points contained in the set $\mathscr{B}(\bar{Y}^{\text{opt}}, r)$, to (approximately) determine whether the solution corresponds to a global maximizer using the result expressed by Property 2. If we can find an element \bar{Y}_i , contained in the set \mathcal{P} , such that $\bar{\mathcal{J}}(\bar{Y}_i) \leq \bar{\mathcal{J}}(\bar{Y}^{\text{opt}})$, then the algorithm is restarted, but now updating the initial guess for the IPM to \bar{Y}_i .

Remark 22: For the nonlinear mapping associated with the CorPower-like device defined in Section V-A, the heuristic

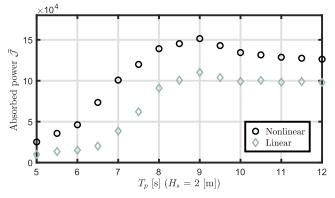


Fig. 5. Constrained (displacement and velocity) power absorption for the nonlinear moment-based energy-maximizing controller proposed in this article (black circles) and its linear counterpart (gray diamonds).

discussed above provides a global solution virtually always after a single iteration (with the cardinality of the set \mathcal{P} chosen as P = 50). In particular, comparisons have been carried out against global optimization routines based on genetic algorithms (GAs), to determine whether the solution obtained with the proposed heuristic effectively coincides with that computed by GA. The IPM utilized to solve (51) is based on [54]. Note that a range of other numerical optimization methods can be equally applied to solve (51).

Remark 23: The moment-based controller normalized run time, i.e., the ratio between the time required to compute the energy-maximizing optimal control input for the duration of the simulation and the length of the simulation itself, is always less than a second for the totality of the preceding simulations, which is consistent with the typical sampling time of a full-scale WEC [15], hence achieving real-time performance. We note that the real-time application of the proposed technique can be performed in a receding-horizon fashion, by directly following the moment-based methodology described in [32, Sec. IV] (see also Remark 4).

We now present performance results for the proposed nonlinear moment-based controller, in terms of energy absorption, under both displacement and velocity constraints. Fig. 5 explicitly shows the value of \mathcal{J} (black circles), for sea states with $H_s = 2$ [m] and $T_p \in [5, 12]$, where the displacement and velocity of the CorPower-like device are constrained to $X_{\text{max}} = 2$ [m] and $X_{\text{max}} = 2$ [m/s], respectively. In addition, Fig. 5 shows the performance of the linear momentbased controller (gray diamonds), i.e., solving the OCP in Corollary 3, assuming that $\overline{\mathcal{J}}_{nl}$ is zero, applied to the nonlinear system described by (14). It can be readily appreciated that the performance of the proposed nonlinear approach outperforms its linear counterpart, for the totality of the sea states analyzed in this study, with differences of up to $\approx 45\%$ in total power absorption. We note that, though not penalized in the results of Fig. 5 (to offer a best case scenario for the linear controller), the solution based on linear assumptions can often violate the physical limitations imposed as state constraints, as a direct consequence of ignoring nonlinear effects in the computation of such an energy-maximizing control law. This is illustrated and discussed in the following paragraph, where we fully expose the capabilities of the nonlinear moment-based control strategy presented in this study.

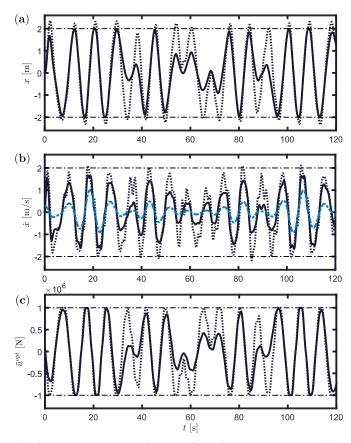


Fig. 6. Motion and control results for polychromatic wave excitation with $H_s = 2$ [m] and $T_p = 10$ [s], for both linear (black dotted line) and nonlinear (black solid line) moment-based controllers. (a) Displacement. (b) Velocity and (scaled) wave excitation force input (blue dashed–dotted line). (c) Corresponding control inputs, used to elicit the motion results. The dashed–dotted horizontal lines represent constraint values.

Fig. 6 shows the time histories of displacement (\mathbf{a}) , velocity (b), and control input (c), for a specific example of sea-state realization with $T_p = 10$ [s], where we have also included a maximum control (PTO) force constraint $U_{\text{max}} = 1 \times 10^6$ [N]. Some key features associated with the presented momentbased strategy can be directly appreciated from Fig. 6, as discussed in the following. To begin with, the state and input limits, under the action of the nonlinear moment-based control strategy (solid black), are being consistently respected throughout the complete simulation, hence illustrating the capability of the approach to maximize energy absorption for WECs subject to nonlinear hydrodynamic effects while respecting the physical limitations of both device and actuator (PTO). This is clearly not the case for the solution based on linear assumptions (black dotted line), where a consistent violation can be appreciated, for both displacement and velocity (state constraints). Though not significant (in magnitude) for this particular sea state, this violation happens consistently in time and can potentially damage device components.

Finally, we note that, as can be appreciated in Fig. $6(\mathbf{b})$, the velocity of the device under optimal control conditions, for both linear and nonlinear moment-based controllers, remains "in-phase¹²" with the (scaled) wave excitation force

¹²We use the term "in-phase" to indicate that the peaks (local maxima and minima) of both signals are aligned in time.

(blue dashed–dotted line), agreeing with well-known theoretical results for unconstrained energy-maximization of WECs [24].

VI. CONCLUSION

This study introduces a nonlinear moment-based energymaximizing control framework for WECs, subject to both state and input constraints. The use of nonlinear moments, in conjunction with an appropriate approximation method (based on the family of weighted residual methods), allows the objective function, associated with the energy-maximizing OCP, to be mapped to a finite-dimensional NP, which can be solved efficiently by state-of-the-art numerical solvers. We guarantee the existence of a globally optimal solution within the presented framework, under mild assumptions. In addition, we give explicit conditions that relate local and global optima, which are effectively exploited in the numerical implementation. The performance of this method is illustrated through a case study, where a CorPower-like device is considered, subject to nonlinear hydrostatic restoring force and viscous forces. We show that physical limitations are consistently respected within this nonlinear moment-based framework, maximizing absorbed energy while effectively minimizing the risk of component damage. Comparisons are presented with its linear counterpart, i.e., a moment-based controller without explicitly considering hydrodynamic nonlinearities when computing the control law, consistently showing improved performance for the totality of the sea states analyzed, with up to $\approx 45\%$ of increase in energy absorption. Finally, future work will compare the presented nonlinear moment-based approach with established techniques in the WEC control field, both in terms of computational speed and energy-maximizing performance.

Appendix Outer Γ-Convexity

To keep this article self-contained, we recall some facts from [43]. These are particularly useful for the results derived in Section IV-B. In particular, we begin by recalling the definition of outer Γ -convexity, originally proposed in [19] and [43].

Definition 5 (Outer Γ -Convexity [19], [43]): Let $\Gamma \subset \mathbb{R}^n$ and $g : \mathbb{R}^n \to \mathbb{R}$. The function g is said to be outer Γ convex on a given nonempty convex set $D \subset \mathbb{R}^n$ if, for all $\{x_0, x_1\} \subset D$ satisfying $x_0 - x_1 \notin \Gamma$, there is a closed subset $Z \subset [0, 1]$ containing $\{0, 1\}$ such that

$$[x_0, x_1] \subset \{(1 - \zeta)x_0 + \zeta x_1 \mid \zeta \in Z\} + \frac{1}{2}\Gamma$$
 (62)

and

$$g((1-\zeta)x_0 + \zeta x_1) \le (1-\zeta)g(x_0) + \zeta g(x_1)$$
(63)

 $\forall \zeta \in Z \setminus \{0, 1\}$. If the inequality (63) holds strictly, then g is said to be strictly outer Γ -convex.

Definition 5 has shown to be useful in deriving specific properties of boundedly perturbed strictly convex quadratic functions [43]. In particular, let $A \in \mathbb{R}^{n \times n}$ be a symmetric positive definite matrix, $b \in \mathbb{R}^n$, and

$$f(x) := x^{\mathsf{T}} A x + b^{\mathsf{T}} x \tag{64}$$

with $x \in \mathbb{R}^n$. For a given set $D \subset \mathbb{R}^n$, (64) gives origin to the convex QP, termed problem (*P*)

$$\min_{x \in D} f(x). \tag{65}$$

Consider now the modified problem (\tilde{P})

$$\min_{x \in D} \tilde{f}(x) := f(x) + p(x) \tag{66}$$

where $p : \mathbb{R}^n \to \mathbb{R}$ is lower semicontinuous. For convenience, we call p the perturbation, \tilde{f} the perturbed function, and (\tilde{P}) the perturbed problem. We now recall the following fundamental result from [43].

Theorem 2 [43, Th. 2.2]: Suppose that the perturbation p is bounded by a parameter s, i.e., $\sup_{x \in D} |p(x)| \le s < +\infty$. Then, the perturbed function $\tilde{f} = f + p$ is strictly outer Γ -convex on D for $\Gamma = \mathscr{B}(0, r)$, with $r = (2s / \min \lambda(A))^{1/2}$.

The result of Theorem 2 shows that a bounded perturbation p does not completely destroy the strict convexity of the quadratic function $f(x) = x^{\mathsf{T}}Ax + b^{\mathsf{T}}x$, but the perturbed function $\tilde{f} = f + p$ is still strictly outer Γ -convex for some suitable set $\Gamma \subset \mathbb{R}^n$. This has two fundamental implications for the perturbed problem (\tilde{P}) , which we recall below as properties. Note that we explicitly state the source of each of these properties, by pointing to specific results of [43]. In what follows, we shall assume that the perturbation p is bounded by a parameter so that Theorem 2 can be used.

Property 3 [43, Lemma 4.3]: Let x_{QP}^* be the global minimizer of problem (P) and suppose that the set $\mathscr{B}(x_{QP}^*, r) \cap D$ is closed. Then, problem (\tilde{P}) always admits a global minimizer.

Property 4 [43, Th. 3.1]: Let x^* be a local Γ -minimizer of \tilde{f} , i.e.,

$$\tilde{f}(x^*) = \min_{x \in \mathscr{B}(x^*, r) \cap D} \tilde{f}(x).$$
(67)

Then, x^* is a global minimizer of \tilde{f} , i.e.,

$$\tilde{f}(x^*) = \min_{x \in D} \tilde{f}(x).$$
(68)

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